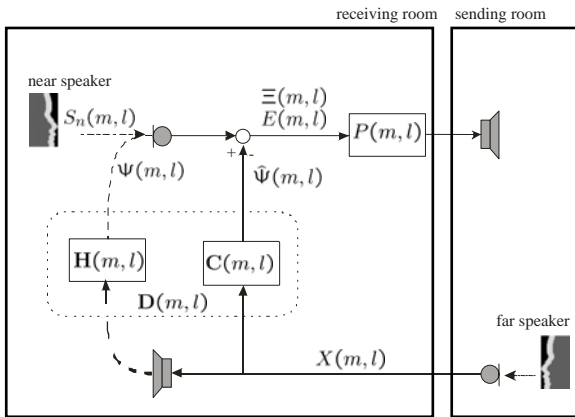


Motivation

- Post-Filtering is a common enhancement technique for the conventional acoustic echo canceller (AEC), especially, when it is diverged or too short.
- A reliable estimate of the (residual) echo power spectral density (PSD) is essential for designing the Post-Filter.
- For short DFT-lengths resulting in low latency a partitioned calculation is necessary, which leads to biased estimates. This bias has to be corrected.

System model

Partitioned frequency domain system model of an acoustic echo canceller in front of a Post-Filter:



$$\begin{aligned} X(m, l) &= [X(m, l) \dots X(m, l - L_H + 1)]^T \\ H(m, l) &= [H_0(m, l) \dots H_{L_H-1}(m, l)]^T \\ C(m, l) &= [C_0(m, l) \dots C_{L_{AEC}-1}(m, l) \quad 0 \dots 0]^T \\ D(m, l) &= H(m, l) - C(m, l) \end{aligned}$$

m : discrete frequency index; $L_{AEC} = L'_{AEC} \cdot L_{DFT}$: length of AEC
 l : discrete block index; $L_H = L'_H \cdot L_{DFT}$: length of room impulse response

Design of the Post-Filter

Spectral subtraction:
$$P(m, l) = \frac{\hat{\Phi}_{EE}(m, l) - \hat{\Phi}_{\Xi\Xi}(m, l)}{\hat{\Phi}_{EE}(m, l)}$$

Calculation of the residual echo:
$$\hat{\Xi}(m, l) = \hat{D}^T(m, l) \cdot X(m, l)$$

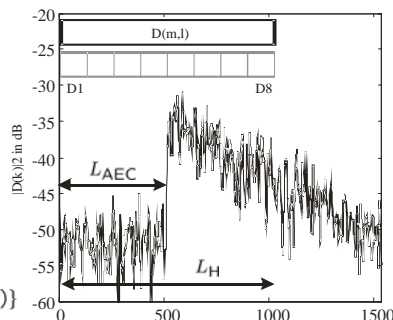
Est. of the system misalignment:
$$\hat{D}(m, l) = R_{XX}^{-1}(m, l) \cdot \hat{\Phi}_{XE}(m, l)$$

Partitioned Calculation of the System Misalignment (I)

Assuming temporarily low correlated loudspeaker signals in the DFT-domain, $E\{X^*(m, l-i)X(m, l-k)\} \approx 0$, $\forall i \neq k$, the partitions of the system misalignment can be calculated independently:

$$D_i(m, l) = \frac{\Phi_{i,X\Xi}(m, l)}{\Phi_{XX}(m, l-i)}$$

$$\Phi_{i,X\Xi}(m, l) = E\{X^*(m, l-i)\Xi(m, l)\}$$



Partitioned Calculation of the System Misalignment (II)

With $\Xi(m, l) = \sum_{i=0}^{L_H-1} D_i(m, l) \cdot X(m, l-i)$ the first partition can be rewritten exemplarily:

$$\begin{aligned} \frac{\Phi_{0,X\Xi}(m, l)}{\Phi_{XX}(m, l)} &= \frac{E\{X^*(m, l) \cdot \sum_{i=0}^{L_H-1} D_i(m, l) \cdot X(m, l-i)\}}{E\{X^*(m, l) \cdot X(m, l)\}} \\ &= \underbrace{D_0(m, l)}_{\text{true value}} + \underbrace{\frac{\sum_{i=1}^{L_H-1} D_i(m, l) \cdot E\{X^*(m, l) \cdot X(m, l-i)\}}{E\{X^*(m, l) \cdot X(m, l)\}}}_{\text{bias}} \end{aligned}$$

After substituting the expectation operators $E\{\cdot\}$ by estimation methods $\hat{E}\{\cdot\}$ in practical realizations, the bias is nonzero.

Optimal Smoothing of the System Misalignment

A first order recursive smoothing of the system misalignment can be exploited especially with nonstationary signals:

$$|\hat{D}_0(m, l)|^2 = \alpha |\hat{D}_0(m, l-1)|^2 + (1-\alpha) \frac{|\hat{\Phi}_{0,X\Xi}(m, l)|^2}{\hat{\Phi}_{XX}(m, l)}$$

An optimal smoothing factor for the mean of a stochastic system $\hat{D}_0(m, l)$ is obtained by minimizing

$$\left(|\hat{D}_0(m, l)|^2 - E\{|\hat{D}_0(m, l)|^2\} \right)^2 \stackrel{!}{=} \min.$$

$$\alpha_{opt}(m, l) = \frac{1}{1 + \frac{E\{|\hat{D}_0(m, l)|^2\} - |\hat{D}_0(m, l-1)|^2}{\sum_{i=1}^{L_H-1} E\{|\hat{D}_i(m, l)|^2\} \Phi_{XX}(m, l-i)}}$$

Practical Realization and Simulation Results

- The factor $|E\{|\hat{D}_0(m, l)|^2\} - |\hat{D}_0(m, l-1)|^2|$ can be interpreted as a permitted adaptation speed and set constant.

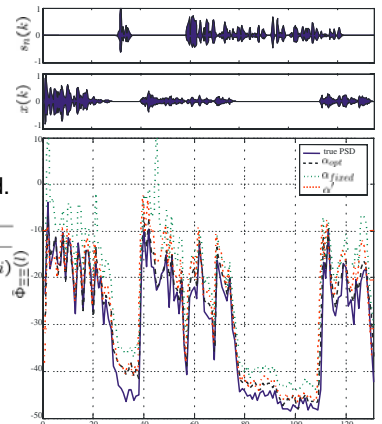
- Typical values for \tilde{C} are in the range of $0.08 < \tilde{C} < 0.12$ and thus we choose $\tilde{C} = 0.1$.

- As a simplification the dependence on the system misalignment can be neglected.

$$\alpha'(m, l) = \frac{1}{1 + C' \frac{\hat{\Phi}_{XX}(m, l)}{\sum_{i=1}^{L_H-1} \hat{\Phi}_{XX}(m, l-i)}}$$

- Fixed Alpha as reference

$$\alpha_{fixed} = e^{-\frac{F_B}{T_s}}$$



Conclusions

- A frequency dependent first order recursive smoothing factor was presented which is robust and straight forward to implement
- Drastical enhancement of the system identification by Wiener-Hopf-Equation
- Reliable redundancy for AEC