

Spatial Sensitivity for Listening Room Compensation

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Motivation

- Listening Room Compensation (LRC) / Room Impulse Response Shaping can increase speech intelligibility.
- Since LRC devices are designed for fixed positions of lousdpeaker and microphone spatial mismatch degrades performance.
- This contribution demonstrates the effects of spatial mismatch visually and acoustically.

Listening Room Compensation

- An equalizer precedes the acoustic channel
- Common design method: Least Squares Equalizer $c_{FQ} = H^+d$
- Problem: Channel h[k] is needed!



• The desired system d[k] is approximated by the overall system of $\mathbf{c}_{\mathsf{EQ}}[k] * h[k]$

Room Impulse Response Shaping

The goal is not spectral flatness of the overall system but a concentration of the energy at a specified temporal envelope (desired area d_d).

 $\mathbf{d}_d = \mathsf{diag}\{\mathbf{w}_d\}\mathbf{Hc}_{\mathsf{EQ}}$

- $d_u = diag\{1 w_d\}Hc_{FO}$
- Maximization of the energy of d_d while keeping the energy of d_u constant leads to the impulse response shortener after Melsa [MYR96, MMEJ03].

 $\mathbf{B}_{\mathsf{BP}} \cdot \mathbf{c}_{\mathsf{EQ},\mathsf{opt}} = \mathbf{A} \cdot \mathbf{c}_{\mathsf{EQ},\mathsf{opt}} \cdot \lambda_{\mathsf{max}}$

$$\begin{split} \mathbf{A} &= \mathbf{H}^{H} \text{diag} \left\{ \mathbf{w}_{\text{BP},\text{d}} \right\}^{2} \mathbf{H} \\ \mathbf{B}_{\text{BP}} &= \mathbf{H}_{\text{BP}}^{H} \text{diag} \left\{ \mathbf{w}_{\text{BP},\text{d}} \right\}^{2} \mathbf{H}_{\text{BP}} \end{split}$$

- Problem: spectral peaks occur in overall transfer function!
- Post processing by a linear prediction filter can reduce the spectral



- · For reducing the problem of late echoes the impulse response should better be shaped than shortened [KM05].
- This can be done by a exponential decreasing window.





- LRC devices are usually designed for fixed and a-priori known positions of source and microphone.
- If the assumptions of the spatial configuration are violated severe distortions may occur.





 Spectral Flatness Measure (SFM) for a LS-EQ, order 2048, $\tau_{60} = 200 \text{ ms}$, position 3

Constrained LS-EQ design

• If an estimation error for the transfer function exist $H = \hat{H} + \tilde{H}$ the least squares EQ can be modified considering these errors.

$$\mathbf{c}_{EQ} = (\hat{\mathbf{H}}^T \hat{\mathbf{H}} + \underbrace{\tilde{\mathbf{U}}_{T}^T \tilde{\mathbf{H}}}_{\approx \delta \cdot ||\hat{\mathbf{h}}||^2 \mathbf{I}})^{-1} \hat{\mathbf{H}} \mathbf{d}$$

$$\overset{6}{=} \underbrace{(\hat{\mathbf{H}}^T \hat{\mathbf{H}} + \underbrace{\tilde{\mathbf{U}}_{T} \tilde{\mathbf{H}}}_{\approx \delta \cdot ||\hat{\mathbf{h}}||^2 \mathbf{I}})^{-1} \hat{\mathbf{H}} \mathbf{d}$$

$$\overset{6}{=} \underbrace{(\hat{\mathbf{H}}^T \hat{\mathbf{H}} + \underbrace{\tilde{\mathbf{U}}_{T} \tilde{\mathbf{H}}}_{\approx \delta \cdot ||\hat{\mathbf{h}}||^2 \mathbf{I}})^{-1} \hat{\mathbf{H}} \mathbf{d}$$

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$$\overset{6}{=} \underbrace{(\hat{\mathbf{H}}^T \hat{\mathbf{H}} + \underbrace{\tilde{\mathbf{U}}_{T} \tilde{\mathbf{H}}}_{\approx \delta \cdot ||\hat{\mathbf{h}}||^2 \mathbf{I}})^{-1} \hat{\mathbf{H}} \mathbf{d}$$

Spectral peaks due to spatial mismatch can be strongly reduced.

Impulse Response Smoothing

Complex Fractional Octave Smoothing as proposed in [HM00] is also capable to increase spatial robustness.



• RIR shaping and constrained LS-EQ design lead to spatially more robust LRC designs than Complex Smoothing.



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Quality Assessment

- Subjective quality assessment is time consuming and expensive.
- Reliable and commonly accepted quality measurement (of enhancement) for dereverberation algorithms has yet to be found.

Objective Quality Measures for LRC

• Several measures can be found in the literature to evaluate dereverberation algorithms:

Measures based on the transfer function:

Variance of logarithmic transfer function [Mou94]:

 $\sigma_{\mathcal{H}}^2 = \frac{1}{n_{max} - n_{min} + 1} \sum_{n=n_{min}}^{n_{max}} (20 \cdot \log_{10} |\mathcal{H}[n]| - \bar{\mathcal{H}}_{dB})^2$ $\bar{\mathcal{H}}_{dB} = \frac{1}{n_{max} - n_{min} + 1} \sum_{n=n_{min}}^{n_{max}} 20 \cdot \log_{10} |\mathcal{H}[n]|$

 $\mathcal{H}[n] = H[n]C_{\mathsf{EQ}}[n]$: concatenated system of EQ and RTF

n : discrete frequency index

 n_{min} : frequency index corresponding to $f=\rm 200 Hz$

 n_{min} : frequency index corresponding to $f=\rm 3600Hz$

Spectral Flatness Measure (SFM), [Joh88]

 $\mathsf{SFM} = \frac{G_n}{A_n} = \frac{\sqrt[N]{\prod_{n=0}^{N-1} |\mathcal{H}[n]|^2}}{\frac{1}{N} \sum_{n=0}^{N-1} |\mathcal{H}[n]|^2}$ G[n] : Geometric mean

A[n]: Arithmetic mean

Measures based on the impulse response: $D_{50} = \frac{\sum_{k=0}^{k_{50}-1} (c_{\mathsf{EQ}}[k] * h[k])^2}{\sum_{k=0}^{L_h} (c_{\mathsf{EQ}}[k] * h[k])^2}$ "Deutlichkeit" (D50), [Kut00] $k_{50} = 50 \text{ms}/f_s$: discrete time index corresponding to a time of 50 ms $\mathrm{CI} = \frac{\sum_{k=0}^{k_{80}} (c_{\mathrm{EQ}}[k] * h[k])^{2}[k]}{\sum_{k=k_{80}}^{L_{h}} (c_{\mathrm{EQ}}[k] * h[k])^{2}[k]}$ Clarity Index (CI), [Kut00] $k_{80} = 80 \text{ms}/f_s : \text{discrete time index corresponding to a time of 80 ms}$ Time (CT), [Kut00] $CT = \frac{\sum_{k=0}^{L_h} k(c_{\text{EQ}}[k] * h[k])^2[k]}{\sum_{k=0}^{L_h} (c_{\text{EQ}}[k] * h[k])^2[k]}$ Central Time (CT), [Kut00] Direct-to-Reverberation-Ratio (DRR), [TS06] $\mathsf{DRR}_{\mathsf{dB}} = 10\log_{10} \frac{(c_{\mathsf{EQ}}[k_{h_{max}}] * h[k_{h_{max}}])^2}{\sum_{k \neq k_{h_{max}}} (c_{\mathsf{EQ}}[k] * h[k])^2}$ $k_{h_{max}}$: index of maximum of h[k]Signal-based measures: Segmental Signal-to-Reverberation Ratio (SSRR), [NG05] $\mathsf{SSRR}_{\mathsf{dB}} = \frac{1}{K/L} \sum_{\ell=0}^{K/L-1} 10 \log_{10} \frac{\sum_{k=0}^{L-1} \hat{y}[\ell L+k]^2}{\sum_{k=0}^{L-1} (\hat{y}[\ell L+k] - y[\ell L+k])^2}$ K :signal length L :block length ℓ :block index $\hat{y}[k]$:signal after desired system $\mathbf{d}[k]$ y[k] :microphone signal Bark Spectral Distortion (BSD), [Yan99] $\mathsf{BSD} = \frac{\frac{1}{K/L} \sum_{\ell=0}^{K/L-1} \sum_{i=1}^{Q} \left[L_{\hat{y}}^{(\ell)}(i) - L_{\hat{y}}^{(\ell)}(i) \right]^2}{\frac{1}{K/L} \sum_{\ell=0}^{K/L-1} \sum_{i=1}^{Q} \left[L_{\hat{y}}^{(\ell)}(i) \right]^2}$ Q: No. of critical bands i: No. of critical band $L_{\hat{u}}(\ell)$: Bark spectrum of frame ℓ $L_y(\ell)$: Bark spectrum of frame ℓ Cepstral Distance (CD), [Yan99] $\mathsf{CD} = \frac{1}{K/L} \sum_{\ell=1}^{K/L-1} \left\{ \frac{10}{\ln 10} \sqrt{ \left[C_{\mathsf{LP}}^{(\hat{y})}(1,\ell) - C_{\mathsf{LP}}^{(y)}(1,\ell) \right]^2 + 2\sum_{i=2}^m \left[C_{\mathsf{LP}}^{(\hat{y})}(i,\ell) - C_{\mathsf{LP}}^{(y)}(i,\ell) \right]^2 } \right\}$ $C_{\mathsf{LP}}^{(x)}(i) = a_i + \frac{1}{i} \sum_{i=1}^{i-1} (i-j) \cdot C_{\mathsf{LP}}^{(x)}(i-j) \cdot a_j \quad \text{for } i = 1, 2, 3...$ $a = -R_{xx}^{-1}r_{xx}$ $C_{LP}^{(y)}$: Cepstral Coefficients for signal y

i : Cepstral Coefficient index **a** : Coefficients of Linear Predictor The different LRC schemes can be compared by means of various objective measures (see left column). Furthermore a subjective quality assessment is possible by applying the precomputed equalizers to a given sound-file and by listening to the equalized sound-file acoustically.

MATLAB® Demo



- Possible EQ designs:
- Possible EQ orders (@8kHz): 128, 256, 51
- Possible room reverberation times: 50ms, 100ms, 200ms, 400ms, 800ms, 1200m

RIR smoothing, RIR shaping 128, 256, 512, 1024, 2048 50ms, 100ms, 200ms, 400ms, 800ms, 1200ms

LS-EQ, Constrained LS-EQ,

Possible spatial positions:



References

- [HM00] P. Hatziantoniou and J.N. Mourjopoulos: "Generalized Fractional-Octave Smoothing of Audio and Acoustic Responses", J. of the Audio Engineering Society, vol. 48, no. 4, pp. 259-280, April 2000
- [Joh88] J. D. Johnston: "Transform Coding of Audio Signals using Perceptual Noise Criteria", IEEE Journal on Selected Areas in Communication, vol. 6, no. 2, pp. 314-232, Feb. 1988
- [KM05] M. Kallinger and A. Mertins: "Room Impulse Response Shortening by Channel Shortening Concepts", Proc. of Asilomar Conference on Signals, Systems, and Computers, Pacific Grove, USA, 2005
- [Kut00] H. Kuttruff: "Room Acoustics", 4th Edition, 2000
- [Mou94] J. N. Mourjopoulos. "Digital Equalization of Room Acoustics", Journal of the Audio Engineering Society, vol. 42, no.11, pp. 884–900, Nov. 1994.

[MYR96] P.J.W. Melsa, R.C. Younce, and C.E. Rohrs: "Impulse Response Shortening for Discrete Multitone Transceivers", IEEE Trans. On Communications, vol. 44, no. 12, pp. 1662-1672, Dec. 1996

- [MMEJ03] R.K. Martin, D. Ming, B.L. Evans, and C.R. Johnson Jr.: "Efficient Channel Shortening Equalizer Design", J. on Applied Signal Processing, vol. 13, pp 1279-1290, Dec. 2003
- [NG05] P.A. Naylor, and N.D. Gaubitch: "Speech Dereverberation", in Proc. of the International Workshop on Acoustic Echo and Noise Control, (IWAENC'05), Eindhoven, The Netherlands, 2005
- [TS06] M. Triki and D.T.M. Slock: "**Iterated Delay and Predict Equalization for Blind Speech Dereverberation**", in Proc. of the International Workshop on Acoustic Echo and Noise Control, (IWAENC'06), Paris, France, Sep. 2006.
- [WSG92] S. Wang and A. Sekey and A. Gersho: "An Objective Measure for Predicting Subjective Quality of Speech Coders", IEEE J. on Selected Areas of Communications, vol. 10, no. 5, pp. 819-829, Feb. 1992

[Yan99] W. Yang, "Enhanced Modified Bark Spectral Distortion (EMBSD): A Objective Speech Quality Measure Based on Audible Distortion and Cognition Model", PhD-thesis, Temple University, Philadelphia, USA, May 1990